

# Tunable Capacitor For Superconducting Qubits Using an InAs/InGaAs Heterostructure

C30: I/O, Packaging, & 3D Integration for Superconducting and Semiconductor Qubits I

Nick Materise<sup>1</sup>, Matthieu Dartiailh<sup>2</sup>, Javad Shabani<sup>2</sup>, Eliot Kapit<sup>1</sup>

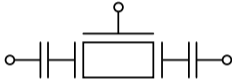
<sup>1</sup>Colorado School of Mines, Department of Physics

<sup>2</sup>Center for Quantum Phenomena, Department of Physics, New York University

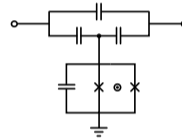
March 15, 2021

## Voltage vs. Flux Biased Couplers

### Voltage-controlled coupler<sup>1</sup>



### Flux-tunable coupler<sup>2</sup>

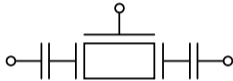


<sup>1</sup>See A37.00009, P29.00013

<sup>2</sup>Adapted from J. Martinis, SQUnT 2020

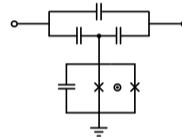
## Voltage vs. Flux Biased Couplers

### Voltage-controlled coupler<sup>1</sup>



- + 2nd order sensitive to voltage bias
- + No heating introduced by voltage bias
- + Minimal cross-talk expected

### Flux-tunable coupler<sup>2</sup>



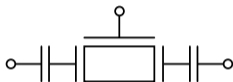
- 1st order sensitive to flux bias
- Current biases can heat system
- Cross-talk from stray magnetic fields

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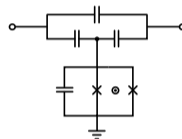
## Voltage vs. Flux Biased Couplers

### Voltage-controlled coupler<sup>1</sup>



- + 2nd order sensitive to voltage bias
- + No heating introduced by voltage bias
- + Minimal cross-talk expected
  - Not part of cQED fab, yet
  - Dielectric loss from 2DEG possible

### Flux-tunable coupler<sup>2</sup>

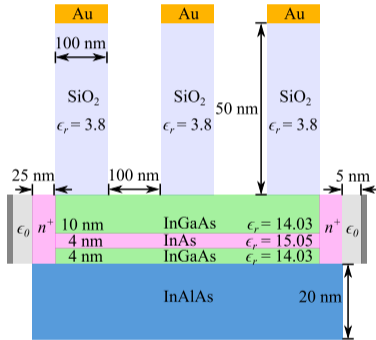


- 1st order sensitive to flux bias
- Current biases can heat system
- Cross-talk from stray magnetic fields
- + Compatible with cQED fab
- + Loss from distributed capacitors

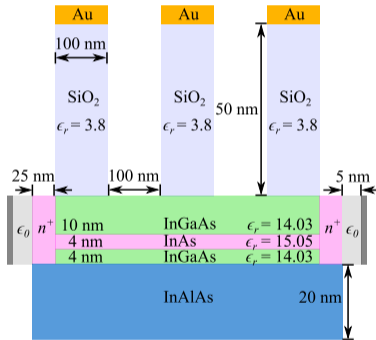
<sup>1</sup>See A37.00009, P29.00013

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## Device Design & Semiconductor Simulations



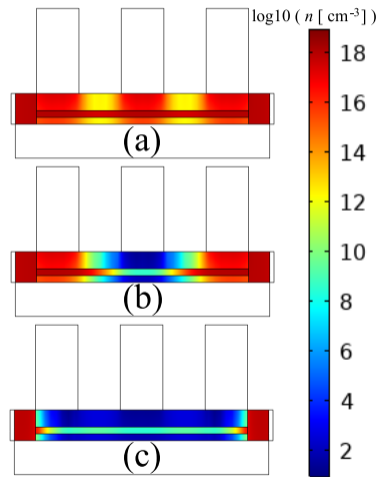
## Device Design & Semiconductor Simulations



- ▶ Electron densities calculated from equilibrium solutions to the drift-diffusion equations with Fermi-Dirac statistics in COMSOL
- ▶ 2DEG confinement included with density gradients modifying the equilibrium electron densities<sup>3</sup>
- ▶ Electrostatic charge conservation applied to InAlAs, SiO<sub>2</sub>, and air-gap dielectric regions

<sup>3</sup>Ancona, Journal of Computational Electronics **10**, 65 (2011).

## Device Design & Semiconductor Simulations

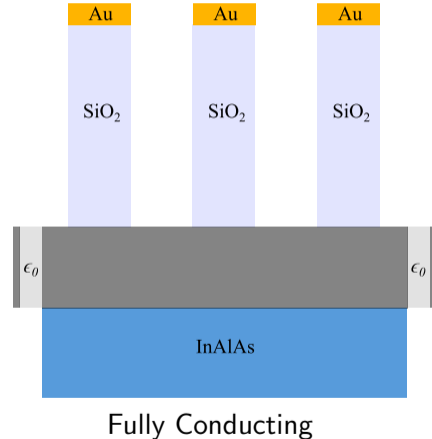


- ▶ Electron densities calculated from equilibrium solutions to the drift-diffusion equations with Fermi-Dirac statistics in COMSOL
- ▶ 2DEG confinement included with density gradients modifying the equilibrium electron densities<sup>3</sup>
- ▶ Electrostatic charge conservation applied to InAlAs, SiO<sub>2</sub>, and air-gap dielectric regions
- ▶ Electron densities for three gate configurations

<sup>3</sup>Ancona, Journal of Computational Electronics 10, 65 (2011).

## Capacitance Matrix DC Calculations

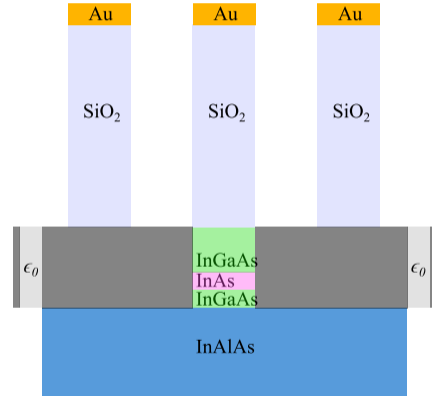
- ▶ Simplify geometry into perfect electric conductors and ideal dielectrics in the fully conducting / depleted limits of the 2DEG
- ▶ End-to-end capacitance limited by air-gap capacitors (enlarged to reduce aspect ratio in COMSOL simulations)





## Capacitance Matrix DC Calculations

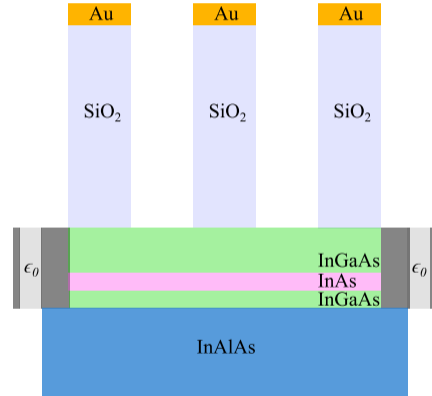
- ▶ Simplify geometry into perfect electric conductors and ideal dielectrics in the fully conducting / depleted limits of the 2DEG
- ▶ End-to-end capacitance limited by air-gap capacitors (enlarged to reduce aspect ratio in COMSOL simulations)



Fully Depleted Single Gate

## Capacitance Matrix DC Calculations

- ▶ Simplify geometry into perfect electric conductors and ideal dielectrics in the fully conducting / depleted limits of the 2DEG
- ▶ End-to-end capacitance limited by air-gap capacitors (enlarged to reduce aspect ratio in COMSOL simulations)



Fully Depleted Three Gates

## Capacitance Matrix DC Calculations

- ▶ Solve Poisson's equation with the charge continuity equation in 2D

$$-\nabla \cdot d(\epsilon_0 \nabla V - P) = \rho$$

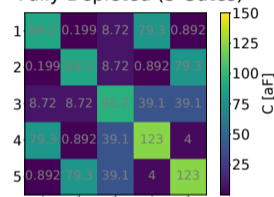
- ▶ Extract Maxwell capacitance matrix

$$\begin{pmatrix} Q_1 \\ Q_2 \\ \vdots \\ Q_N \end{pmatrix} = \begin{pmatrix} C_{11} & C_{12} & \dots & C_{1N} \\ C_{21} & C_{22} & \dots & C_{2N} \\ \vdots & \vdots & & \vdots \\ C_{N1} & C_{N2} & \dots & C_{NN} \end{pmatrix} \begin{pmatrix} V_1 \\ V_2 \\ \vdots \\ V_N \end{pmatrix}$$

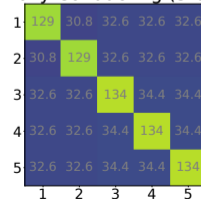
- ▶ On/off ratio of

$$C_{12}^{\text{on}} / C_{12}^{\text{off}} = 30.8 \text{ aF} / 0.199 \text{ aF} \approx 160!$$

Fully Depleted (3 Gates)



Fully Conducting (3 Gates)



## Conductance Matrix Calculations

- Solve time harmonic equations

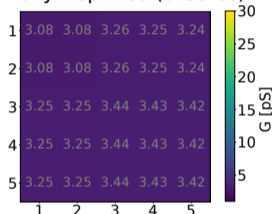
$$\begin{aligned}\nabla \cdot (\sigma \mathbf{E} + \mathbf{J}_e) + i\omega\rho &= 0, \\ \nabla \cdot \mathbf{D} &= \rho\end{aligned}$$

- Use terminal current, voltage solutions to compute admittance matrix,  $\mathbf{Y}$

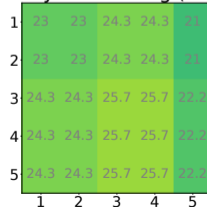
$$\begin{pmatrix} V_1 \\ V_2 \\ \vdots \\ V_N \end{pmatrix} = \begin{pmatrix} Y_{11} & Y_{12} & \dots & Y_{1N} \\ Y_{21} & Y_{22} & \dots & Y_{2N} \\ \vdots & \vdots & & \vdots \\ Y_{N1} & Y_{N2} & \dots & Y_{NN} \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \\ \vdots \\ I_N \end{pmatrix}$$

$$\mathbf{Y} = \mathbf{G} + i\omega\mathbf{C}$$

Fully Depleted (3 Gates)



Fully Conducting (3 Gates)



## Dielectric Loss Estimation

- ▶ Estimation of dielectric loss-limited  $T_1$ <sup>4,5</sup>

$$T_1^{-1} = \frac{\omega}{Q} = \omega \sum_j \frac{p_j}{Q_j} + \Gamma_0, \quad Q_j^{-1} = \tan \delta_j$$

$$p_j = \frac{t_j \varepsilon_{1,j} \int_{S_j} |E|^2 dS}{\int_V |E|^2 dV}$$

	$t_j$ [nm]	$p_j$	$\tan \delta_j$	$T_1$ [ $\mu$ s]
InGaAs	10	2.08E-5	4.8E-5	3190
InAs	4	3.18E-5	4.8E-5	20800
InGaAs	4	2.86E-5	4.8E-5	23200
InAlAs	20	5.64E-4	4.8E-5	1180
SiO <sub>2</sub> <sup>7</sup>	50	4.44E-3	2.00E-3	3.58
Total	-	-	-	3.57

- ▶ III-V loss tangents approximated with GaAs<sup>6</sup>
- ▶ Limited by SiO<sub>2</sub>, pending measurement of III-Vs

<sup>4</sup>Wenner et al., Applied Physics Letters **99**, 113513 (2011).

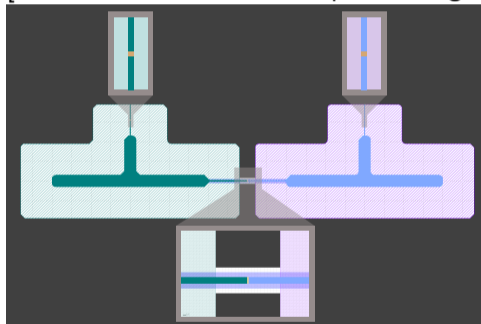
<sup>5</sup>Wang et al., Applied Physics Letters **107**, 162601 (2015).

<sup>6</sup>McRae et al., arXiv e-prints, arXiv:2009.10101 (2020).

<sup>7</sup>Li et al., IEEE Transactions on Applied Superconductivity **23**, 1501204 (2013).

## Matrix Element Extraction – Energy Participation Ratios

Two Transmon + Coupler HFSS Model  
[Insets: Junctions and Coupler Enlarged]



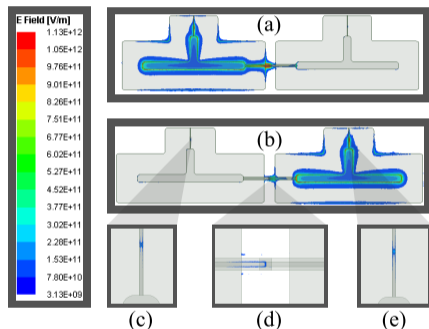
- ▶ HFSS Eigenmodes  $\rightarrow$  EPR  $\rightarrow$  Matrix Elements<sup>8,9</sup>
- ▶ Josephson junction inductances, capacitances as inputs
- ▶ Coupler modeled as lumped element capacitor

<sup>8</sup>Mineev, arXiv e-prints, arXiv:1902.10355 (2019).

<sup>9</sup>Mineev et al., arXiv e-prints, arXiv:2010.00620 (2020).

# Matrix Element Extraction – Energy Participation Ratios

## Two Transmon + Coupler Eigenmodes



**Figure:** Electric field norm of first three eigenmode solutions. (a) 5.9 GHz, (b) 6.1 GHz, and (c-e) 8.6 GHz; scaled views of (c) qubit 1, (d) coupling element, and (e) qubit 2 electric field norms of third eigenmode.

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## Matrix Element Extraction – Energy Participation Ratios

Mode No.	$\omega/2\pi$ [GHz]	Q	$\chi/2\pi$ [MHz]		
1 (d)	5.667	4.5E8	<b>226</b>	62.5	0.965
2 (d)	5.838	1.3E9	62.5	<b>226</b>	1.11
3 (d)	8.614	1.8E13	0.965	1.11	<b>0.002</b>
1 (c)	5.669	4.5E8	<b>223</b>	67.1	0.974
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**Table:** Extracted matrix elements from energy participation ratio calculations for values of  $R_{12}$ ,  $C_{12}$  in the fully depleted (d) and fully conducting (c) limits of the 2DEG. Diagonal entries of  $\chi$  are scaled by 1/2 to denote the anharmonicities.

- ▶ HFSS Eigenmodes  $\rightarrow$  EPR  $\rightarrow$  Matrix Elements<sup>8,9</sup>
- ▶ Josephson junction inductances, capacitances as inputs
- ▶ Coupler modeled as lumped element capacitor
- ▶ Parametric sweep over the coupler capacitance to extract self- and cross-Kerr matrix elements

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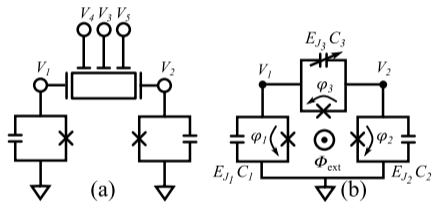
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- ▶ Parametric sweep over the coupler capacitance to extract self- and cross-Kerr matrix elements
- ▶ Retain anharmonicities for exchange interaction matrix elements

<sup>8</sup>Mineev, arXiv e-prints, arXiv:1902.10355 (2019).

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## Matrix Element Extraction – Exchange Interaction

- ▶ Extract exchange  $Q_i Q_j$  interaction matrix elements  $C^{-1}/2$  from<sup>10,11</sup>



$$H = \frac{1}{2} \mathbf{Q}^T \mathbf{C}^{-1} \mathbf{Q} + \sum_j E_{J_j} (1 - \cos \varphi_j)$$

$$\mathbf{C} = \begin{pmatrix} C_1 + C_3 & -C_3 \\ -C_3 & C_2 + C_3 \end{pmatrix}$$

$$C_k = \frac{e^2}{2E_C} \simeq -\frac{e^2}{2\alpha_k} = -\frac{e^2}{2\chi_{kk}}$$

<sup>10</sup>Orlando et al., Phys. Rev. B **60**, 15398 (1999).

<sup>11</sup>Koch et al., Phys. Rev. A **76**, 042319 (2007).

## Matrix Element Extraction – Exchange Interaction

- ▶ Extract exchange  $Q_i Q_j$  interaction matrix elements  $C^{-1}/2$  from<sup>10,11</sup>

Qubit Index	Matrix Elements	[MHz]
1 (d)	226	<b>0.0005</b>
2 (d)	0.0005	226
1 (c)	226	<b>0.079</b>
2 (c)	0.079	226

**Table:** Charge-charge (exchange) interaction matrix elements in the (d) depleting and (c) conducting limits of the 2DEG coupler.

$$H = \frac{1}{2} \mathbf{Q}^T C^{-1} \mathbf{Q} + \sum_j E_{J_j} (1 - \cos \varphi_j)$$

$$C = \begin{pmatrix} C_1 + C_3 & -C_3 \\ -C_3 & C_2 + C_3 \end{pmatrix}$$

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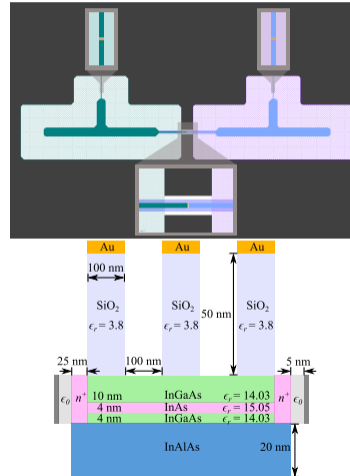
- ▶ Exchange interaction matrix elements recover  $\approx 160$  on/off ratio

<sup>10</sup>Orlando et al., Phys. Rev. B **60**, 15398 (1999).

<sup>11</sup>Koch et al., Phys. Rev. A **76**, 042319 (2007).

# Summary

- ▶ Capacitive couplers are drop-in replacements for their inductive counterparts
- ▶ We estimate on/off ratios of two orders of magnitude in two transmon geometry
- ▶ Dielectric loss dominated by SiO<sub>2</sub>, pending experimental verification
- ▶ Opens the door for future semiconductor-superconducting hybrid systems



## Acknowledgements

- ▶ We acknowledge support from the Graduate Fellowships for STEM Diversity, NSF grant No. PHY-1653820, ARO grant No. W911NF-18-1-0125 and LPS/ARO grant No. W911NF-18-1-0115
- ▶ We would like to thank Joel Howard, Bradley Lloyd, Paul Niyonkuru, David Rodríguez Pérez, Zhijie Tang, Meenakshi Singh, William Strickland, and Joseph Yuan for helpful discussions

## References

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